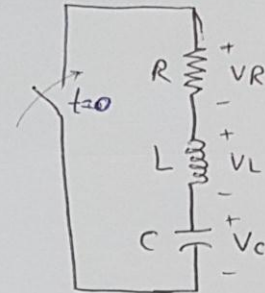


6- RLC circuit.

The series RLC circuit shown in fig(24) contains no source & KVL for closed loop when switch is closed at $t=0$



fig(24)

$$V_R + V_L + V_C = 0$$

$$Ri + L \frac{di}{dt} + \frac{1}{C} \int i dt = 0$$

differentiating and dividing by L

$$\frac{d^2 i}{dt^2} + \frac{R}{L} \frac{di}{dt} + \frac{1}{LC} i = 0$$

It is important that this same (homogeneous D.E) its like of this expression

$$\ddot{i} + \frac{R}{L} \dot{i} + \frac{1}{LC} i = 0$$

then the solution of this D.E is

$$i = A_1 e^{s_1 t} + A_2 e^{s_2 t}$$

$$A_1 e^{s_1 t} \left(s_1^2 + \frac{R}{L} s_1 + \frac{1}{LC} \right) + A_2 e^{s_2 t} \left(s_2^2 + \frac{R}{L} s_2 + \frac{1}{LC} \right) = 0$$

that is if s_1 and s_2 are the two roots of $(s^2 + \frac{R}{L} s + \frac{1}{LC})$

$$\Rightarrow s_1 = -\frac{R}{2L} + \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}} = -\alpha + \beta$$

$$s_2 = -\frac{R}{2L} - \sqrt{\left(\frac{R}{2L}\right)^2 - \frac{1}{LC}} = -\alpha - \beta$$

where $\alpha = R/2L$; $\omega_0 = \frac{1}{\sqrt{LC}}$; $\beta = \sqrt{\alpha^2 - \omega_0^2}$

(16)

Now there are three cases

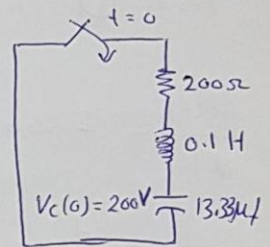
- ① overdamped case ($\alpha > \omega_0$)
- ② critically damped case ($\alpha = \omega_0$)
- ③ underdamped case ($\alpha < \omega_0$)

Now we will discuss all of these cases

① overdamped case ($\alpha > \omega_0$)

here $i = e^{-\alpha t} (A_1 e^{\beta t} + A_2 e^{-\beta t})$ where α and β are real positive numbers

Ex: - The circuit as shown in fig(25) obtain the current transient if switch is closed at $t=0$



fig(25)

Solution

for this circuit.

$$\alpha = \frac{R}{2L} = 10^3 \text{ s}^{-1}$$

$$\omega_0^2 = \frac{1}{Lc} = 7.5 \times 10^5 \text{ s}^{-2}$$

$$\beta = \sqrt{\alpha^2 - \omega_0^2} = 500 \text{ s}^{-1}$$

$$i = e^{-1000t} (A_1 e^{500t} + A_2 e^{-500t})$$

to find the constant A_1 & A_2

KVL at $t=0$

$$\overset{\text{zero}}{(0)}R + L \frac{di}{dt} + V_c(0) = 0$$

$$\frac{di}{dt} = \frac{-V_c(0)}{L} = \frac{-200}{0.1} = -2000 \text{ A/s}$$

at $t=0$

$$i = A_1 + A_2 = 0$$

$$\text{and } \frac{di}{dt} = -2000 = -500 A_1 - 1500 A_2 \Rightarrow A_1 = -2 \text{ \& } A_2 = 2$$

$$\Rightarrow i = -2 e^{-500t} + 2 e^{-1500t}$$

(17)

⑫ critically damped case ($\alpha = \omega_0$)

when $\alpha = \omega_0$

$$\Rightarrow \beta = 0 \quad \text{i.e. } S_1 = S_2$$

the solution of D.E becomes

$$i = e^{-\alpha t} (A_1 + A_2 t)$$

Ex₁: - the same circuit as shown in fig (25) but the value of $C = 10 \mu\text{F}$

$$\alpha = \frac{R}{2L} = 10^3 \text{ s}^{-1} \quad \omega_0 = \frac{1}{\sqrt{LC}} = 10^3 \text{ s}^{-2}$$

$$\beta = 0$$

$$i = e^{-1000t} (A_1 + A_2 t)$$

$$i(0) = 0 \quad \left. \frac{di}{dt} \right|_0 = -2000 \text{ A/s}$$

$$i(0) = 0 \Rightarrow A_1 = 0$$

$$\text{and } A_2 = -2000$$

$$i(t) = -2000 t e^{-1000t} \text{ A}$$

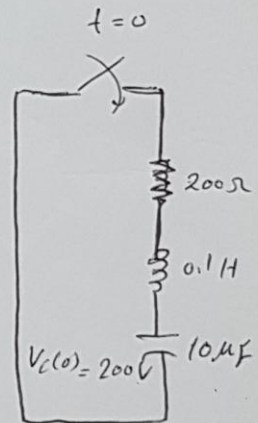
$$i(0) = 0 = A_1 \Rightarrow A_1 = 0$$

$$\left. \frac{di}{dt} \right|_0 = -2000 = \frac{-1000}{1} e^{-1000 \cdot 0} (A_1 + A_2 \cdot 0)$$

$$\Rightarrow -2000 = -1000 e^{-1000 \cdot 0} + A_2 t \cdot (-1000) e^{-1000 \cdot 0} + e^{-1000 \cdot 0} \cdot A_2$$

$$-2000 = A_2$$

$$i(t) = -2000 t e^{-1000t}$$



(18)

③ underdamped case ($\alpha < \omega_0$)

In this case β is a pure imaginary part $\beta = j|\beta|$
 then the solution of $D_1 E$ becomes

$$i = e^{-\alpha t} (A_1 \cos |\beta|t + A_2 \sin |\beta|t)$$

OR

$$i = e^{-\alpha t} A_3 \sin (|\beta|t + \phi)$$

where A_3 and ϕ are two new constant.

Ex - the same circuit as shown in fig (25) but the value of $C = 1 \mu f$

$$i(0) = 0 \quad \left. \frac{di}{dt} \right|_0 = -2000$$

$$\alpha = 1000 \text{ s}^{-1}$$

$$\beta = \sqrt{10^6 - 10^7} = j3000$$

$$i = e^{-1000t} A_3 \sin(3000t + \phi)$$

when $t = 0$ $i = 0$ and $\phi = 0$ and $A_3 = -0.667 \text{ A}$

$$i = -0.667 e^{-1000t} \sin 3000t \text{ (A)}$$

hint:- $i = e^{-\alpha t} (A_1 \cos |\beta|t + A_2 \sin |\beta|t) \quad \dots \text{ ①}$

$$i = A_3 \sin (\beta t + \phi)$$

by derivative of eq. ①

$$= e^{-\alpha t} (k_1 e^{j\beta t} + k_2 e^{-j\beta t})$$

$$= e^{-\alpha t} (k_1 (\cos \beta t + j \sin \beta t) + k_2 (\cos \beta t - j \sin \beta t))$$

$$= e^{-\alpha t} ((k_1 + k_2) \cos \beta t + j(k_1 - k_2) \sin \beta t)$$

$$= e^{-\alpha t} (A_1 \cos |\beta|t + A_2 \sin |\beta|t) \quad ; \quad A_1 = k_1 + k_2 \quad ; \quad A_2 = j(k_1 - k_2)$$

(19)

7 - Two mesh circuit:-

from fig (26) we get

$$R_1 i_1 + L_1 \frac{di_1}{dt} + R_1 i_2 = V \quad \text{--- (1)}$$

$$R_1 i_1 + (R_1 + R_2) i_2 + L_2 \frac{di_2}{dt} = V \quad \text{--- (2)}$$

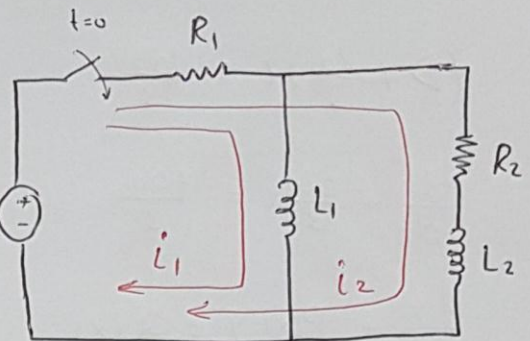


fig (26)

derivative eq. (1) we get

$$R_1 \frac{di_1}{dt} + L_1 \frac{d^2 i_1}{dt^2} + R_1 \frac{di_2}{dt} = 0 \quad \text{--- (3)}$$

from eq. (2)

$$\frac{di_2}{dt} = \frac{V - (R_1 i_1 + (R_1 + R_2) i_2)}{L_2} \quad \text{--- (4)}$$

from eq. (1)

$$i_2 = \frac{V - (R_1 i_1 + L_1 \frac{di_1}{dt})}{R_1} \quad \text{--- (5)}$$

sub. eq. (4) & eq. (5) in to eq. (3) and rearrangement, we get

$$\frac{d^2 i_1}{dt^2} + \left(\frac{R_1 L_1 + R_2 L_1 + R_1 L_2}{L_1 L_2} \right) \frac{di_1}{dt} + \frac{R_1 R_2}{L_1 L_2} i_1 = \frac{V R_2}{L_1 L_2} \quad \text{--- (6)}$$

the steady-state solution of eq. (6) ; $i_1(\infty) = \frac{V}{R_1}$

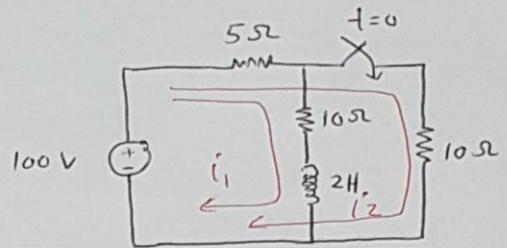
$$S^2 + \left(\frac{R_1 L_1 + R_2 L_1 + R_1 L_2}{L_1 L_2} \right) S + \frac{R_1 R_2}{L_1 L_2} = 0$$

$$i(0) = 0 \quad ; \quad \frac{di_1}{dt} = \frac{V}{L_1}$$

Ex1- write the D.E for circuit (27) and solve for i_1 and i_2 when the switch is closed at $t=0$

$$15i_1 + 2 \frac{di_1}{dt} + 5i_2 = 100 \quad \text{--- (1)}$$

$$5i_1 + 15i_2 = 100 \quad \text{--- (2)}$$



from eq. (2) $i_2 = \frac{100 - 5i_1}{15} = \frac{20 - i_1}{3}$ --- (3) fig(27)

sub eq. (3) in (1) we get

$$15i_1 + 2 \frac{di_1}{dt} + \frac{100 - 5i_1}{3} = 100$$

$$45i_1 + 6 \frac{di_1}{dt} + 100 - 5i_1 = 300$$

$$\frac{6 di_1}{dt} + 40i_1 = 200$$

$$\frac{di_1}{dt} + \frac{20}{3}i_1 = \frac{100}{3}$$

$$(D + \frac{20}{3})i_1 = \frac{100}{3} \quad \text{--- (4)}$$

solution of eq. (4) is

$$i_1 = C e^{-\frac{20}{3}t} + \frac{\frac{100}{3}}{\frac{20}{3}} = C e^{-\frac{20}{3}t} + 5 \quad \text{--- (5)}$$

at $t=0$ $i_1 = \frac{100}{15} = \frac{20}{3}$ (from eq. (4)) --- (6)

sub eq. (6) in eq. (5) we get

$$\frac{20}{3} = C + 5 \Rightarrow C = \frac{20}{3} - 5 = \frac{5}{3}$$

$$i_1 = \frac{5}{3} e^{-\frac{20}{3}t} + 5 \quad \text{--- (7)}$$

sub eq. (7) in (3) we get

$$i_2 = \frac{20}{3} - \frac{5}{9} e^{-\frac{20}{3}t} - \frac{5}{3} = 5 - \frac{5}{9} e^{-\frac{20}{3}t}$$

(21)

Tutorial

Ex 1 In the fig(28), the switch is closed at $t=0$, when the 6Mf capacitor has charge $Q_0 = 300\text{ }\mu\text{C}$. Obtain the expression for the transient voltage V_R

Solution:-

$$C_{eq} = C_1 // C_2 + C_0$$

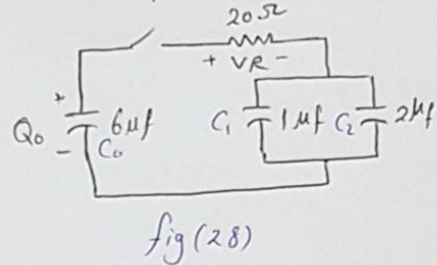
$$= \frac{(C_1 + C_2) \times C_0}{(C_1 + C_2) + C_0} = \frac{3 \times 6}{3 + 6} = 2\text{ }\mu\text{f}$$

$$\Rightarrow \tau = RC_{eq} = 20 \times 2 = 40\text{ }\mu\text{s}$$

At $t=0$ KVL gives

$$V_R = \frac{Q_0}{C_0} = \frac{300}{6} = 50\text{ V} = V_0$$

$$\therefore V(t) = V_0 e^{-\frac{t}{\tau}} = 50 e^{-\frac{t}{40}} \text{ V if } t \text{ in } \mu\text{s}$$



Ex:- In the fig(29) the switch is moved from position ① to position ② at $t=0$, Obtain the current i_2 at $t = 34.7\text{ ms}$.

Solution:-

when the switch at the position ②

$$L_{eq} = L_1 + L_2 // L_3$$

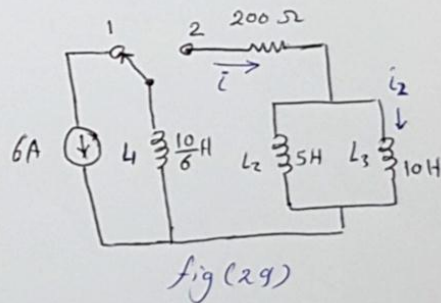
$$= \frac{10}{6} + \frac{5 \times 10}{5 + 10} = 5\text{ H}$$

$$\Rightarrow \tau = \frac{L_{eq}}{R} = \frac{5}{200} = 25\text{ ms}$$

$$\Rightarrow i = i_0 e^{-\frac{t}{\tau}} = 6 e^{-\frac{t}{25}} \text{ A}$$

$$i_2 = \frac{5}{15} i = \frac{5}{15} \times 6 e^{-\frac{t}{25}} = 2 e^{-\frac{t}{25}}$$

$$i_2(34.7) = 2 e^{-\frac{34.7}{25}} = 0.5\text{ A}$$



(22)

Ex 2 In fig (30) the switch is closed at $t=0$, obtain the current i and capacitor voltage V_c for $t > 0$

Solution:-

for $t > 0$.

$$R_{eq} = R_1 // R_2 = \frac{10 \times 10}{10 + 10} = 5 \Omega$$

$$\tau = R_{eq} \cdot C = 5 \times 2 = 10 \text{ } \mu\text{s}$$

at $t \rightarrow \infty$ then the capacitor is open

$$i(\infty) = \frac{50}{20} = 2.5 \text{ A}$$

$$V_c(\infty) = i \times R_2 = 2.5 \times 10 = 25 \text{ V}$$

$$V_c(t) = [V_c(0) - V_c(\infty)] e^{-\frac{t}{\tau}} + V_c(\infty)$$

$$= 25(1 - e^{-\frac{t}{10}}) \text{ V} \quad \text{if } t \text{ in } \mu\text{s}.$$

$$i_c = C \frac{dV}{dt} = 5 e^{-\frac{t}{10}} \text{ A}$$

the current in the parallel R_2 is

$$i_{10\Omega} = \frac{V_c}{10} = 2.5(1 - e^{-\frac{t}{10}}) \text{ A}$$

hence:

$$i = i_c + i_{10\Omega} = 2.5(1 + e^{-\frac{t}{10}}) \text{ A}.$$

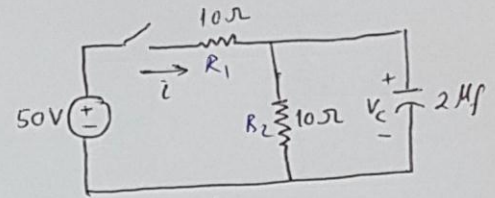


fig (30)

(23)

Ex 8 for the fig (31), obtain the current i_L for all values of t

Solution- for $t < 0$ 50V source is 50V

$$\bar{i} = \frac{50}{R_1 + R_2} = \frac{50}{20} = 2.5 \text{ A}$$

Now for $t > 0$

and $t \rightarrow \infty$ this current divides equally between the two 10Ω resistor.

$i_L(\infty) = -2.5 \text{ A}$. the time constant of the circuit (τ)

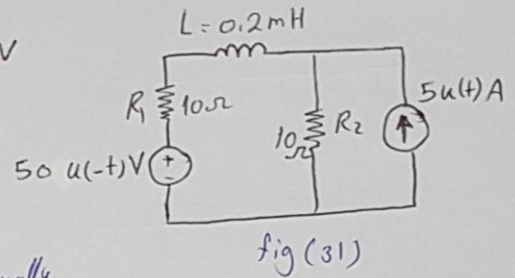
$$\tau = \frac{0.2 \times 10^{-3}}{R_1 + R_2} = \frac{0.2 \times 10^{-3}}{20} = \frac{1}{100} \text{ ms.}$$

and so, with t in ms and using $i_L(0^+) = i_L(0^-) = 2.5 \text{ A}$

$$\begin{aligned} i_L &= i_L(\infty) + [i_L(0) - i_L(\infty)] e^{-\frac{t}{\tau}} \\ &= 5 e^{-100t} - 2.5 \text{ (A)} \end{aligned}$$

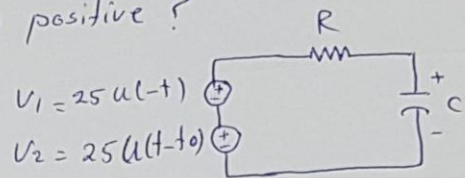
and finally, using unit step functions to combine the expression for $t < 0$ and $t > 0$

$$i_L = 2.5 u(-t) + (5 e^{-100t} - 2.5 u(t)) \text{ A}$$



Ex 41 - A series RC circuit with $R = 5 \text{ K}\Omega$ and $C = 20 \mu\text{f}$, has two voltage sources in series: $V_1 = 25 u(-t) \text{ V}$ & $V_2 = 25 u(t-t_0) \text{ V}$. Obtain the complete expression for the voltage across the capacitor and sketch it if t_0 is positive?

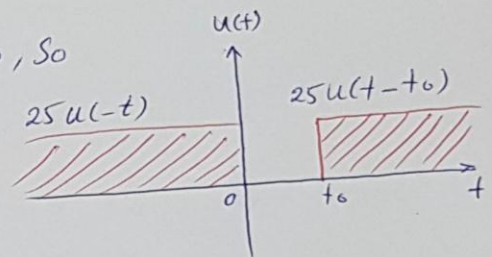
Solution:-



① for $t \leq 0$; $V_C = 25 \text{ V}$

② for $0 \leq t \leq t_0$ both sources are zero, so that V_C decays exponentially from 25 V towards zero $\tau = 5 \times 10^3 \times 20 \times 10^{-6} = 100 \times 10^{-3} = 0.1$

$$V_C = 25 e^{-\frac{t}{\tau}} = 25 e^{-10t} \text{ (V)}$$



③ for $t \geq t_0$

$$V_C(t_0) = 25 e^{-10t_0}$$

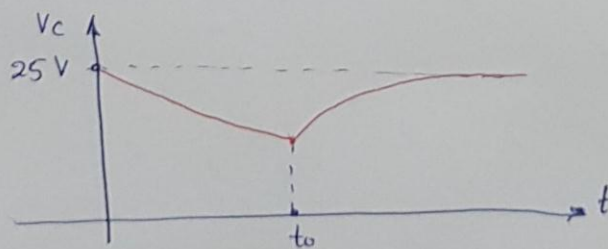
the V_C builds from $V_C(t_0)$ to the final value of 25 V established by V_2

$$V_C = [V_C(t_0) - V_C(\infty)] e^{-\frac{(t-t_0)}{\tau}} + V_C(\infty)$$

$$= 25 [1 - (e^{+10t} - 1) e^{-10t}] \text{ (V) for all } t \geq t_0$$

thus for t

$$V_C(t) = 25 u(-t) + 25 e^{-10t} [u(t) - u(t-t_0)] + 25 [1 - (e^{10t} - 1) e^{-10t}] u(t-t_0) \text{ V}$$



(25)

Ex 51 - A series RLC circuit with $R = 50 \Omega$; $L = 0.1 \text{ H}$ and $C = 50 \mu\text{F}$, has a constant voltage $V = 100 \text{ V}$ applied at $t = 0$ obtain the current transient, assuming zero initial charge on the capacitor

Solution:-

$$\alpha = \frac{R}{2L} = \frac{50}{2 \times 0.1} = 250 \text{ s}^{-1}$$

$$\omega_0^2 = \frac{1}{LC} = \frac{1}{0.1 \times 50 \times 10^{-6}} = 2 \times 10^5 \text{ s}^{-2}$$

$$\beta = \sqrt{\alpha^2 - \omega_0^2} = \sqrt{(250)^2 - 2 \times 10^5} = j 370.8 \text{ rad/s}$$

\Rightarrow This oscillatory case ($\alpha < \omega_0$) and the general expression is

$$i = e^{-250t} (A_1 \cos 370.8t + A_2 \sin 370.8t)$$

initial condition, obtained is

$$i(0) = 0 ;$$

$$\left. \frac{di}{dt} \right|_0 = 1000 \text{ A/s}$$

then $A_1 = 0$; $A_2 = 2.70 \text{ A}$

$$i = e^{-250t} (2.70 \sin 370.8t)$$

(26)

Ex6:- the switch in the two-mesh circuit shown in fig (32) is closed at $t=0$, obtain the currents i_1 and i_2 for $t > 0$

Solution:-

Method (1) two-mesh.

$$100 = 15i_1 + 10i_2 + 0.01 \frac{di_1}{dt} \quad \text{--- (1)}$$

$$100 = 15i_2 + 10i_1 \quad \text{--- (2)}$$

from eq. (2)

$$i_2 = \frac{100 - 10i_1}{15} \quad \text{--- (3)}$$

sub eq. (3) in (1) we get

$$\frac{di_1}{dt} + 833i_1 = 3333$$

$$i_1 = A e^{-833t} + 4 \text{ A}$$

$$i_1(0) = 0 \Rightarrow A = -4$$

$$i_1 = 4(1 - e^{-833t}) \text{ A}$$

and

$$i_2 = 4 + 2.67 e^{-833t} \text{ A}$$

Method (2)

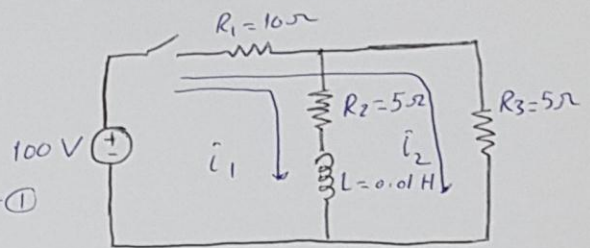
$$R_{eq} = 5 + \frac{5 \times 10}{15} = 8.33 \Omega$$

$$\tau = \frac{L}{R_{eq}} = \frac{0.01}{8.33} = \frac{1}{833} \text{ s}$$

$$\text{at } t = \infty \quad R_T = 10 + \frac{5 \times 5}{5+5} = 12.5 \Omega$$

$$i_T = \frac{100}{12.5} = 8 \text{ A}$$

$$i_L = i_1 = 4(1 - e^{-833t})$$



(27)