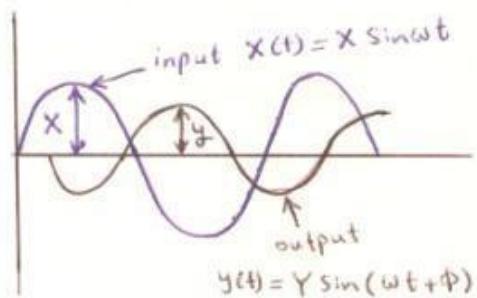
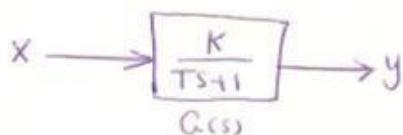


(69)

Frequency - Response Analysis

The term frequency response means the steady-state response of a system to a sinusoidal input, i.e., we vary the frequency of the input signal over a certain range and study the resulting frequency response.

Consider the following system



If $x(t) = X \sin(\omega t)$ then $y(t)$ can be found as follows:

Substituting $j\omega$ for s in $G(s)$ yields

$$G(j\omega) = \frac{K}{j\omega T + 1}$$

$$|G(j\omega)| = \frac{K}{\sqrt{1 + T^2 \omega^2}}$$

while the phase angle ϕ is

$$\phi = \angle G(j\omega) = -\tan^{-1} \omega T$$

$$\text{Thus } y(t) = \frac{XK}{\sqrt{1 + T^2 \omega^2}} \sin(\omega t - \tan^{-1} \omega T)$$

This system is a phase-lag system for large ω .

(70)

Ex:- Consider the network given by

$$G(s) = \frac{s + \frac{1}{T_1}}{s + \frac{1}{T_2}}$$

Determine whether the network is a Lead or Lag network.

Solution: $x(t) = X \sin \omega t$

$$G(j\omega) = \frac{j\omega + \frac{1}{T_1}}{j\omega + \frac{1}{T_2}} = \frac{T_2(1 + T_1 j\omega)}{T_1(1 + T_2 j\omega)}$$

we have

$$|G(j\omega)| = \frac{T_2 \sqrt{1 + T_1^2 \omega^2}}{T_1 \sqrt{1 + T_2^2 \omega^2}}$$

and

$$\phi = \angle G(j\omega) = \tan^{-1} T_1 \omega - \tan^{-1} T_2 \omega$$

$$y(t) = \frac{X T_2 \sqrt{1 + T_1^2 \omega^2}}{T_1 \sqrt{1 + T_2^2 \omega^2}} \sin(\omega t + \tan^{-1} T_1 \omega - \tan^{-1} T_2 \omega)$$

if $T_1 > T_2$, then $\tan^{-1} T_1 \omega - \tan^{-1} T_2 \omega > 0$.

Thus if $T_1 > T_2$

then the network is a Lead network.

If $T_1 < T_2$

then the network is a Lag network.

Frequency-Response characteristics in graphical forms

- ① Nyquist plot or polar plot.
- ② Bode diagram or Logarithmic plot.
- ③ Log-magnitude - versus phase plot (Nichols plots).

polar plot (Nyquist plot)

It is a plot of magnitude of $G(j\omega)$ varies with the phase angle of $G(j\omega)$ on a polar coordinate as ω varied from $0 \rightarrow \infty$. It is often called the Nyquist plot.

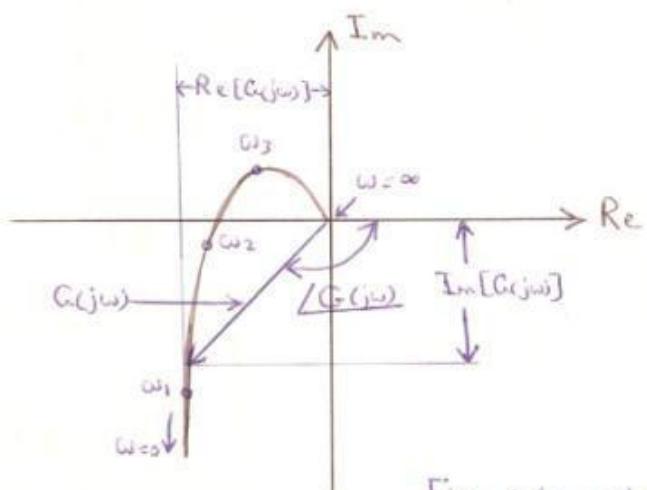


Fig. polar plot

Information about stability is available directly from a polar plot of the sine response of open Loop transfer function $G(j\omega)H(j\omega)$.

(72)

* Integral and derivative factors $(j\omega)^{\pm 1}$

$$G(j\omega) = \frac{1}{j\omega} \quad \text{is the negative imaginary axis.}$$

$$G(j\omega) = \frac{1}{j\omega} = -j \frac{1}{\omega} = \frac{1}{\omega} \angle -90^\circ$$

The polar plot of $G(j\omega) = j\omega$ is the positive imaginary axis.* First-order factors $(1+j\omega T)^{\pm 1}$

$$G(j\omega) = \frac{1}{1+j\omega T} = \frac{1}{\sqrt{1+\omega^2 T^2}} \angle -\tan^{-1}\omega T$$

$$G(j0) = 1 \angle 0^\circ \quad \text{at } \omega = 0$$

$$G(j\frac{1}{T}) = \frac{1}{\sqrt{2}} \angle -45^\circ \quad \text{at } \omega = \frac{1}{T}$$

$$G(j\omega) = X + jY$$

Where $X = \frac{1}{1+\omega^2 T^2}$ = real part of $G(j\omega)$. $Y = \frac{-\omega T}{1+\omega^2 T^2}$ = imaginary part of $G(j\omega)$.

Then we obtain

$$(X - \frac{1}{2})^2 + Y^2 = \left(\frac{1 - \omega^2 T^2}{1 + \omega^2 T^2} \right)^2 + \left(\frac{-\omega T}{1 + \omega^2 T^2} \right)^2 = \left(\frac{1}{2} \right)^2$$

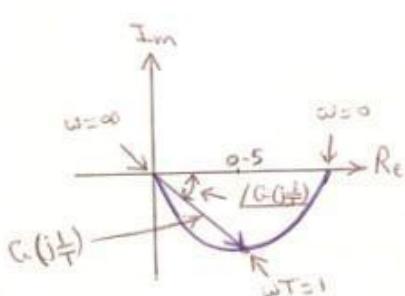


Fig. Polar plot of

$$\frac{1}{1+j\omega T}$$

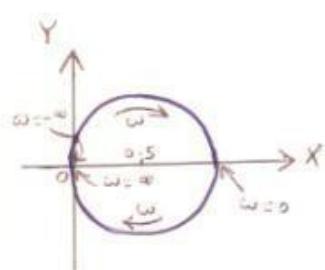
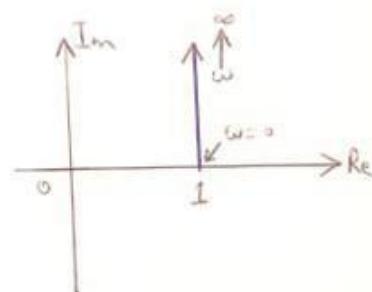
Fig. plot of $G(j\omega)$
in X-Y plane.

Fig. polar plot of

$$1+j\omega T$$

(73)

* Quadratic Factors $[1 + 2\zeta(j\omega/\omega_n) + (j\omega/\omega_n)^2]^{-1}$

$$G(j\omega) = \frac{1}{1 + 2\zeta(j\omega/\omega_n) + (j\omega/\omega_n)^2} \quad \text{for } \zeta > 0$$

$$\lim_{\omega \rightarrow 0} G(j\omega) = 1 \angle 0^\circ \quad ; \quad \lim_{\omega \rightarrow \infty} G(j\omega) = \infty \angle 180^\circ$$

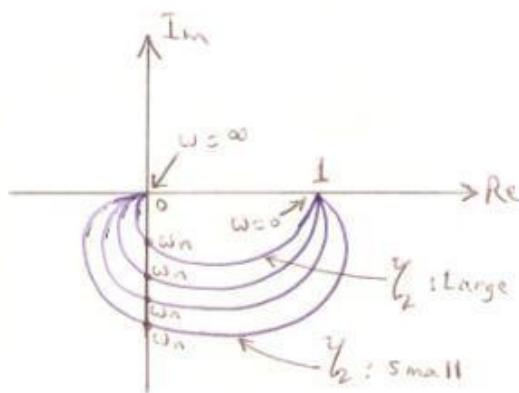


Fig. polar plot of
 $\frac{1}{1 + 2\zeta(j\omega/\omega_n) + (j\omega/\omega_n)^2}$; for $\zeta > 0$

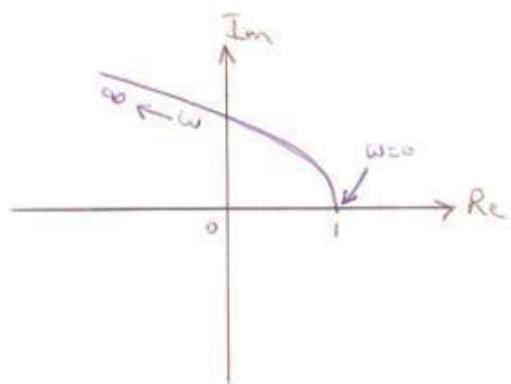


Fig. polar plot of
 $1 + 2\zeta(j\omega/\omega_n) + (j\omega/\omega_n)^2$; for $\zeta > 0$

For $G(j\omega) = 1 + 2\zeta(j\omega/\omega_n) + (j\omega/\omega_n)^2$

$$= (1 - \frac{\omega^2}{\omega_n^2}) + j(\frac{2\zeta\omega}{\omega_n})$$

The low-frequency portion of the curve is

$$\lim_{\omega \rightarrow 0} G(j\omega) = 1 \angle 0^\circ$$

and the high-frequency is

$$\lim_{\omega \rightarrow \infty} G(j\omega) = \infty \angle 180^\circ$$

(74)

Ex 1: Sketch a polar plot of the T.F

$$G(s) = \frac{1}{s(Ts+1)}$$

Solution: $G(j\omega) = \frac{1}{j\omega(1+j\omega T)}$

$$= -\frac{T}{1+\omega^2 T^2} - j \frac{1}{\omega(1+\omega^2 T^2)}$$

$$\lim_{\omega \rightarrow 0} G(j\omega) = -T - j\infty = \infty \angle -90^\circ$$

$$\lim_{\omega \rightarrow \infty} G(j\omega) = 0 - j0 = 0 \angle -180^\circ$$

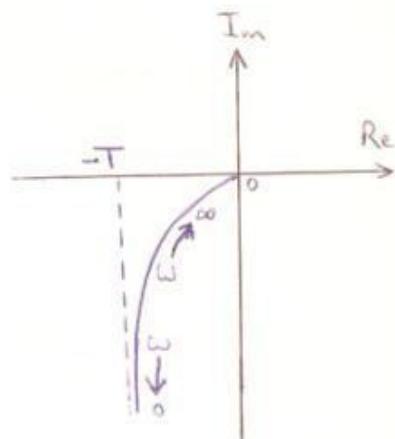


Fig. polar plot of

$$\frac{1}{j\omega(1+j\omega T)}$$

Ex 2: Sketch a polar plot of the T.F

$$G(j\omega) = \frac{e^{-j\omega L}}{1+j\omega T}$$

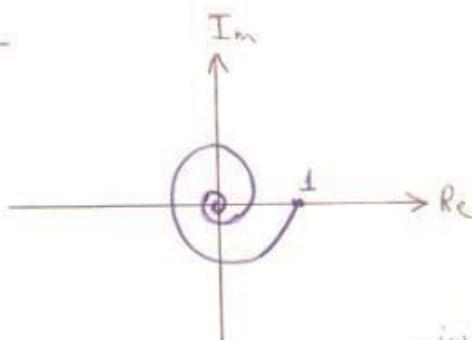
Solution: $G(j\omega) = (e^{-j\omega L}) \left(\frac{1}{1+j\omega T} \right)$

$$|G(j\omega)| = |e^{-j\omega L}| \cdot \left| \frac{1}{1+j\omega T} \right|$$

$$= \frac{1}{\sqrt{1+\omega^2 T^2}}$$

$$\angle G(j\omega) = \angle e^{-j\omega L} + \angle \frac{1}{1+j\omega T}$$

$$= -\omega L + \tan^{-1} \omega T$$

Fig. polar plot of $\frac{e^{-j\omega L}}{1+j\omega T}$